

±2.5g - 10g Three Axis Low-g Micromachined Accelerometer

The MMA7261Q low cost capacitive micromachined accelerometer features signal conditioning, a 1-pole low pass filter, temperature compensation and g-Select which allows for the selection among 4 sensitivities. Zero-g offset full scale span and filter cut-off are factory set and require no external devices. Includes a Sleep Mode that makes it ideal for handheld battery powered electronics.

Features

- Selectable Sensitivity (2.5g/3.3g/6.7g/10g)
- Low Current Consumption: 500 μA
- Sleep Mode: 3 μA
- Low Voltage Operation: 2.2 V 3.6 V
- 6mm x 6mm x 1.45mm QFN
- · Fast Turn On Time
- High Sensitivity (2.5 g)
- · Integral Signal Conditioning with Low Pass Filter
- · Robust Design, High Shocks Survivability
- · Environmentally Preferred Package
- · Low Cost

Typical Applications

- HDD MP3 Player: Freefall Detection
- · Laptop PC: Freefall Detection, Anti-Theft
- · Cell Phone: Image Stability, Text Scroll, Motion Dialing, E-Compass
- Pedometer: Motion Sensing
- PDA: Text Scroll
- Navigation and Dead Reckoning: E-Compass Tilt Compensation
- · Gaming: Tilt and Motion Sensing, Event Recorder
- Robotics: Motion Sensing

ORDERING INFORMATION				
Device Name	Device Name Temperature Package Range Drawing		Package	
MMA7261Q	– 20 to +85°C	98ASA10651D	QFN-16, Tube	
MMA7261QR2	– 20 to +85°C	98ASA10651D	QFN-16,Tape & Reel	

MMA7261Q

MMA7261Q: XYZ AXIS ACCELEROMETER ±2.5g/3.3g/6.7g/10g

Bottom View



16 LEAD QFN98ASA10651D

Top View

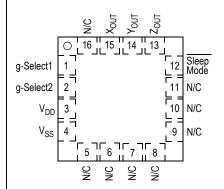


Figure 1. Pin Connections



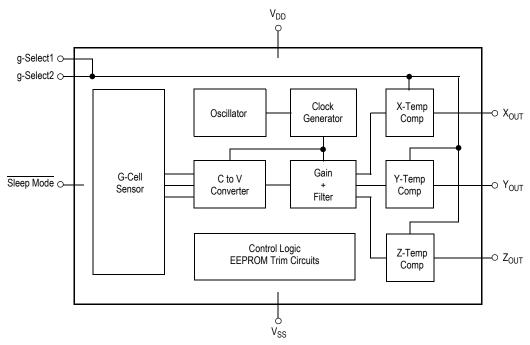


Figure 2. Simplified Accelerometer Functional Block Diagram

Table 1. Maximum Ratings

(Maximum ratings are the limits to which the device can be exposed without causing permanent damage.)

Rating	Symbol	Value	Unit
Maximum Acceleration (all axis)	9 _{max}	±2000	g
Supply Voltage	V_{DD}	-0.3 to +3.6	V
Drop Test ⁽¹⁾	D _{drop}	1.8	m
Storage Temperature Range	T _{stg}	-40 to +125	°C

^{1.} Dropped onto concrete surface from any axis.

ELECTRO STATIC DISCHARGE (ESD)

WARNING: This device is sensitive to electrostatic discharge.

Although the Freescale accelerometer contains internal 2000 V ESD protection circuitry, extra precaution must be taken by the user to protect the chip from ESD. A charge of over 2000 volts can accumulate on the human body or associated test equipment. A charge of this magnitude can

alter the performance or cause failure of the chip. When handling the accelerometer, proper ESD precautions should be followed to avoid exposing the device to discharges which may be detrimental to its performance.

Table 2. Operating Characteristics

Unless otherwise noted: $-20^{\circ}\text{C} \le \text{T}_{\text{A}} \le 85^{\circ}\text{C}$, 2.2 V \le V_{DD} \le 3.6 V, Acceleration = 0g, Loaded output⁽¹⁾

Power Spectral Density RMS $(0.1 \text{Hz} - 1 \text{kHz})^{(4)}$ n_{PSD} — 350 — $\mu g/\sqrt{\text{Hz}}$ Control Timing Power-Up Response Time ⁽⁶⁾ $t_{RESPONSE}$ — 1.0 2.0 ms Enable Response Time ⁽⁷⁾ t_{ENABLE} — 0.5 2.0 ms Sensing Element Resonant Frequency t_{ENABLE} — 0.5 2.0 ms t_{ENABLE} — 0.5 2	Characteristic	Symbol	Min	Тур	Max	Unit
Supply Voltage ⁽³⁾	Operating Range ⁽²⁾					
Supply Current Supply Current at Sieep Mode ⁽⁴⁾ Supply Cur		V _{DD}	2.2	3.3	3.6	V
Supply Current at Sleep Mode(4)			_	500	800	μΑ
Operating Temperature Range T _A -20 — +85 "C Acceleration Range, X-Axis, Y-Axis, Z-Axis g-Select1 & 2: 00 g-Fs — ±2.5 — g g-Select1 & 2: 01 g-Fs — ±3.3 — g g-Select1 & 2: 01 g-Fs — ±6.7 — g g-Select1 & 2: 11 g-Fs — ±10.0 — g Output Signal Zero g (T _A = 25°C, V _{DD} = 3.3 V) ⁽⁵⁾ V _{OFF} 1.485 1.65 1.815 V Zero g (T _A = 25°C, V _{DD} = 3.3 V) V _{OFF} 1.485 1.65 1.815 V Zero g (T _A = 25°C, V _{DD} = 3.3 V) S2.5g 444 480 516 mV/g Sensitivity (T _A = 25°C, V _{DD} = 3.3 V) S2.5g 444 480 516 mV/g 3.3g 3.3g 33.3g 333 360 387 mV/g 9 sensitivity S.Ta 11 120 129 mV/g Sensitivity S.Ta — ±0.03	Supply Current at Sleep Mode ⁽⁴⁾		_	3	10	μA
Acceleration Range, X-Axis, Y-Axis, Z-Axis g-Select 1 & 2: 00 g-Select 1 & 2: 10 g-Select 1 & 2: 11 g-Select 1 & 2: 10 g-Select 1 & 2: 11 g-Select 1 & 2: 10 g-Selec	Operating Temperature Range		-20	_	+85	°C
g-Select1 & 2: 10 gFS — ±3.3 — g g-Select1 & 2: 01 gFS — ±6.7 — g g-Select1 & 2: 11 gFS — ±10.0 — g Output Signal Zero g (TA = 25°C, VDD = 3.3 V) VOFF 1.485 1.65 1.815 V Zero g (TA = 25°C, VDD = 3.3 V) Sensitivity (TA = 25°C, VDD = 3.3 V) Sensit	Acceleration Range, X-Axis, Y-Axis, Z-Axis					
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g-Select1 & 2: 11 gFS — ±10.0 — g Output Signal Zero g (T _A = 25°C, V _{DD} = 3.3 V) ⁽⁵⁾ V _{OFF} 1.485 1.65 1.815 V Zero g (T _A = 25°C, V _{DD} = 3.3 V) V _{OFF} , T _A — ±2 — mg/°C Sensitivity (T _A = 25°C, V _{DD} = 3.3 V) Y _{OFF} , T _A — ±2 — mg/°C Sensitivity (T _A = 25°C, V _{DD} = 3.3 V) Y _{OFF} , T _A — ±2 — mg/°C Sensitivity (T _A = 25°C, V _{DD} = 3.3 V) Y _{OFF} , T _A — ±2 — mg/°C Sensitivity (T _A = 25°C, V _{DD} = 3.3 V) Y _{OFF} , T _A — ±2 — mg/°C Sensitivity (T _A = 25°C, V _{DD} = 3.3 V) Y _{OFF} , T _A — ±2 — mg/°C Sensitivity (T _A = 25°C, V _{DD} = 3.3 V) Y _{OFF} , T _A — ±2 — mg/°C Mry/g Sensitivity (T _A = 25°C, V _{DD} = 3.3 V) Y _{OFF} , T _A — ±2 — mg/°C Mry/g Mry/g Sensitivity (T _A = 25°C, V _{DD} = 3.3 V) 360 387 mr/yg Mry/g M	g-Select1 & 2: 10	9 _{FS}	_	±3.3	_	g
Output Signal Zero g (T _A = 25°C, V _{DD} = 3.3 V) ⁽⁵⁾ V _{OFF} 1.485 1.65 1.815 V Zero g V _{OFF} , T _A — ±2 — mg/°C Sensitivity (T _A = 25°C, V _{DD} = 3.3 V) S.5g 444 480 516 mV/g 3.3g S.3.3g 333 360 387 mV/g 6.7g 10g S.6,7g 167 180 193 mV/g 10g S.10g 111 120 129 mV/g Sensitivity S.7A — ±0.03 — %°°C Bandwidth Response Y 1,3dB — 350 — Hz Hz N°°C Noise RMS (0.1 Hz – 1 kHz) ⁽⁴⁾ n _{RMS} — 4.7 — mVrms N°°C mVrms — Hz N°°C MVrms M°°C	g-Select1 & 2: 01	9 _{FS}	_	±6.7	_	g
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	g-Select1 & 2: 11	-	_	±10.0	_	g
	Output Signal					
	Zero g (T _A = 25°C, V _{DD} = 3.3 V) ⁽⁵⁾	V _{OFF}	1.485	1.65	1.815	V
Sensitivity (T _A = 25°C, V _{DD} = 3.3 V) 2.5g Square Squa			_	±2	_	mg/°C
2.5g	Sensitivity (T _A = 25°C, V _{DD} = 3.3 V)					
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$		S _{2.5a}	444	480	516	mV/g
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	3.3g		333	360	387	mV/g
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	6.7g		167	180	193	mV/g
Sensitivity S,TA — ±0.03 — %°C Bandwidth Response XY $f_{.3dB}$ — 350 — Hz XY $f_{.3dB}$ — 150 — Hz Noise RMS (0.1 Hz – 1 kHz) ⁽⁴⁾ n _{RMS} — 4.7 — mVrms Power Spectral Density RMS (0.1 Hz – 1 kHz) ⁽⁴⁾ n _{PSD} — 350 — μg/./Hz Control Timing t _{RESPONSE} — 1.0 2.0 ms Enable Response Time ⁽⁶⁾ t _{RESPONSE} — 1.0 2.0 ms Sensing Element Resonant Frequency t _{ENABLE} — 0.5 2.0 ms Sensing Sampling Frequency f _{GCELL} — 6.0 — kHz Internal Sampling Frequency f _{CLK} — 11 — kHz Output Stage Performance Full-Scale Output Range (I _{OUT} = 30 μA) V _{FSO} V _{SS} +0.25 — V _{DD} -0.25 V Nonlinearity, X _{OUT} , Y _{OUT} , Z _{OUT} NL _{OUT}	10g		111	120	129	mV/g
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	Sensitivity		_	±0.03	_	%/°C
	Bandwidth Response					
Noise $ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	XY	f _{-3dB}	_	350	_	Hz
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	Z	f _{-3dB}	_	150	_	Hz
Power Spectral Density RMS $(0.1 \ Hz - 1 \ kHz)^{(4)}$ n_{PSD} — 350 — $\mu g/\sqrt{Hz}$ Control Timing Power-Up Response Time ⁽⁶⁾ $t_{RESPONSE}$ — 1.0 2.0 ms Enable Response Time ⁽⁷⁾ t_{ENABLE} — 0.5 2.0 ms Sensing Element Resonant Frequency xY f_{GCELL} — 6.0 — kHz z f_{GCELL} — 3.4 — kHz Internal Sampling Frequency t_{ENABLE} — t_{ENA	Noise					
Control Timing Power-Up Response Time ⁽⁶⁾ Enable Response Time ⁽⁷⁾ Sensing Element Resonant Frequency XY Z Internal Sampling Frequency Z Cutput Stage Performance Full-Scale Output Range ($I_{OUT} = 30 \mu A$) Nonlinearity, X_{OUT} , Y_{OUT} , Z_{OUT} Z Z Z Z Z Z Z	RMS (0.1 Hz – 1 kHz) ⁽⁴⁾	n _{RMS}	_	4.7	_	mVrms
Power-Up Response Time ⁽⁶⁾ $ \begin{array}{ccccccccccccccccccccccccccccccccccc$	Power Spectral Density RMS (0.1 Hz – 1 kHz) ⁽⁴⁾	n _{PSD}	_	350	_	μg/√Hz
Enable Response Time ⁽⁷⁾ $Sensing Element Resonant Frequency$ XY Z $Internal Sampling Frequency$ F_{GCELL} F_{GCELL} F_{CLK} $-$ 11 $-$ KHz $Cutput Stage Performance Full-Scale Output Range (I_{OUT} = 30 \ \mu A) V_{FSO} V_{SS}+0.25 - V_{DD}-0.25 V NL_{OUT} -1.0 - +1.0 \% FSO$	Control Timing					
Sensing Element Resonant Frequency $ \begin{array}{ccccccccccccccccccccccccccccccccccc$		t _{RESPONSE}	_	1.0	2.0	ms
XY f_{GCELL} $ 6.0$ $ kHz$ Z f_{GCELL} $ 3.4$ $ kHz$ Internal Sampling Frequency f_{CLK} $ 11$ $ kHz$ Output Stage Performance Full-Scale Output Range ($I_{OUT} = 30 \mu A$) V_{FSO} $V_{SS} + 0.25$ $ V_{DD} - 0.25$ V Nonlinearity, X_{OUT} , Y_{OUT} , Z_{OUT} V_{COUT}	Enable Response Time ⁽⁷⁾	t _{ENABLE}	_	0.5	2.0	ms
Z Internal Sampling Frequency f_{GCELL} $-$ 3.4 $-$ kHz Internal Sampling Frequency f_{CLK} $-$ 11 $-$ kHz Output Stage Performance Full-Scale Output Range (I _{OUT} = 30 μA) V_{FSO} V_{SS} +0.25 $ V_{DD}$ -0.25 V Nonlinearity, X_{OUT} , Y_{OUT} , Z_{OUT} V_{DUT} $V_$	Sensing Element Resonant Frequency					
Z Internal Sampling Frequency f_{CLK} $-$ 3.4 $-$ kHz f _{CLK} $-$ 11 $-$ kHz Output Stage Performance Full-Scale Output Range (I _{OUT} = 30 μA) V_{FSO} V_{SS} +0.25 $ V_{DD}$ -0.25 V Nonlinearity, X_{OUT} , Y_{OUT} , Z_{OUT} V_{DUT} V_{DU	XY	f _{GCELL}	_	6.0	_	kHz
Output Stage Performance Full-Scale Output Range ($I_{OUT} = 30 \mu A$) V _{FSO} V _{SS} +0.25 V _{DD} -0.25 V Nonlinearity, X _{OUT} , Y _{OUT} , Z _{OUT} NL _{OUT} -1.0 +1.0 %FSO	Z		_	3.4	_	kHz
Full-Scale Output Range ($I_{OUT} = 30 \mu A$) V_{FSO} $V_{SS} + 0.25$ — $V_{DD} - 0.25$ V Nonlinearity, X_{OUT} , Y_{OUT} , Y_{OUT} , Y_{OUT} — $Y_{DD} - 0.25$ V Y_{D	Internal Sampling Frequency	f _{CLK}	_	11	_	kHz
Nonlinearity, X _{OUT} , Y _{OUT} , Z _{OUT} NL _{OUT} -1.0 +1.0 %FSO	Output Stage Performance					
	Full-Scale Output Range (I _{OUT} = 30 μA)	V _{FSO}	V _{SS} +0.25		V _{DD} -0.25	V
Cross Avis Consistivity (8)	Nonlinearity, X _{OUT} , Y _{OUT} , Z _{OUT}	NL _{OUT}	-1.0	_	+1.0	%FSO
Closs-Axis Sensitivity: $V_{XY, XZ, YZ} 5.0 - \%$	Cross-Axis Sensitivity ⁽⁸⁾	V _{XY, XZ, YZ}	_	_	5.0	%

- 1. For a loaded output, the measurements are observed after an RC filter consisting of a 1.0 kΩ resistor and a 0.1 μF capacitor on V_{DD}-GND.
- 2. These limits define the range of operation for which the part will meet specification.
- 3. Within the supply range of 2.2 and 3.6 V, the device operates as a fully calibrated linear accelerometer. Beyond these supply limits the device may operate as a linear device but is not guaranteed to be in calibration.
- 4. This value is measured with g-Select in 2.5g mode.
- 5. The device can measure both + and acceleration. With no input acceleration the output is at midsupply. For positive acceleration the output will increase above $V_{DD}/2$. For negative acceleration, the output will decrease below $V_{DD}/2$.
- 6. The response time between 10% of full scale Vdd input voltage and 90% of the final operating output voltage.
- 7. The response time between 10% of full scale Sleep Mode input voltage and 90% of the final operating output voltage.
- 8. A measure of the device's ability to reject an acceleration applied 90° from the true axis of sensitivity.

MMA7261Q

PRINCIPLE OF OPERATION

The Freescale accelerometer is a surface-micromachined integrated-circuit accelerometer.

The device consists of two surface micromachined capacitive sensing cells (g-cell) and a signal conditioning ASIC contained in a single integrated circuit package. The sensing elements are sealed hermetically at the wafer level using a bulk micromachined cap wafer.

The g-cell is a mechanical structure formed from semiconductor materials (postillion) using semiconductor processes (masking and etching). It can be modeled as a set of beams attached to a movable central mass that move between fixed beams. The movable beams can be deflected from their rest position by subjecting the system to an acceleration (Figure 3).

As the beams attached to the central mass move, the distance from them to the fixed beams on one side will increase by the same amount that the distance to the fixed beams on the other side decreases. The change in distance is a measure of acceleration.

The g-cell beams form two back-to-back capacitors (Figure 3). As the center beam moves with acceleration, the distance between the beams changes and each capacitor's value will change, (C = $A\epsilon/D$). Where A is the area of the beam, ϵ is the dielectric constant, and D is the distance between the beams.

The ASIC uses switched capacitor techniques to measure the g-cell capacitors and extract the acceleration data from the difference between the two capacitors. The ASIC also signal conditions and filters (switched capacitor) the signal, providing a high level output voltage that is ratiometric and proportional to acceleration.

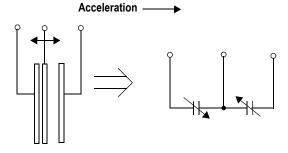


Figure 3. Simplified Transducer Physical Model

SPECIAL FEATURES

g-Select

The g-Select feature allows for the selection among 4 sensitivities present in the device. Depending on the logic input placed on pins 1 and 2, the device internal gain will be changed allowing it to function with a 2.5g, 3.3g, 6.7g, or 10g sensitivity (Table 3). This feature is ideal when a product has applications requiring different sensitivities for optimum performance. The sensitivity can be changed at anytime during the operation of the product. The g-Select1 and g-Select2 pins can be left unconnected for applications requiring only a 2.5g sensitivity as the device has an internal pull-down to keep it at that sensitivity (480mV/g).

Table 3. g-Select pin Descriptions

g-Select2	g-Select1	g-Range	Sensitivity
0	0	2.5g	480mV/g
0	1	3.3g	360mV/g
1	0	6.7g	180mV/g
1	1	10g	120mV/g

Sleep Mode

The 3 axis accelerometer provides a Sleep Mode that is ideal for battery operated products. When Sleep Mode is active, the device outputs are turned off, providing significant reduction of operating current. A low input signal on pin 12 (Sleep Mode) will place the device in this mode and reduce the current to 3uA typ. For lower power consumption, it is recommended to set g-Select1 and g-Select2 to 2.5g mode. By placing a high input signal on pin 12, the device will resume to normal mode of operation.

Filtering

The 3 axis accelerometer contains onboard single-pole switched capacitor filters. Because the filter is realized using switched capacitor techniques, there is no requirement for external passive components (resistors and capacitors) to set the cut-off frequency.

Ratiometricity

Ratiometricity simply means the output offset voltage and sensitivity will scale linearly with applied supply voltage. That is, as supply voltage is increased, the sensitivity and offset increase linearly; as supply voltage decreases, offset and sensitivity decrease linearly. This is a key feature when interfacing to a microcontroller or an A/D converter because it provides system level cancellation of supply induced errors in the analog to digital conversion process.

BASIC CONNECTIONS

Pin Descriptions

Figure 4. Pinout Description

Table 4. Pin Descriptions

Pin No.	Pin Name	Description
1	g-Select1	Logic input pin to select g level.
2	g-Select2	Logic input pin to select g level.
3	V_{DD}	Power Supply Input
4	V_{SS}	Power Supply Ground
5 - 7	N/C	No internal connection. Leave unconnected.
8 - 11	N/C	Unused for factory trim. Leave unconnected.
12	Sleep Mode	Logic input pin to enable product or Sleep Mode.
13	Z _{OUT}	Z direction output voltage.
14	Y _{OUT}	Y direction output voltage.
15	X _{OUT}	X direction output voltage.
16	N/C	No internal connection. Leave unconnected.

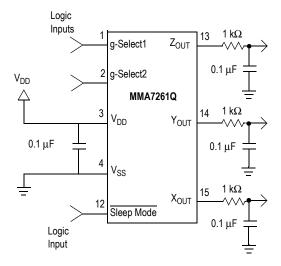


Figure 5. Accelerometer with Recommended Connection Diagram

PCB Layout

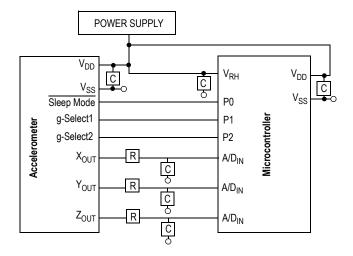
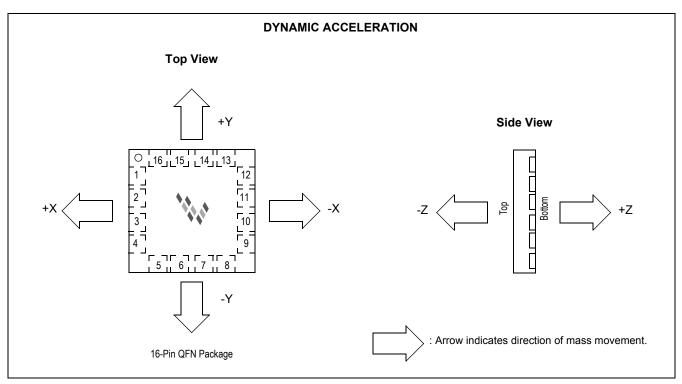


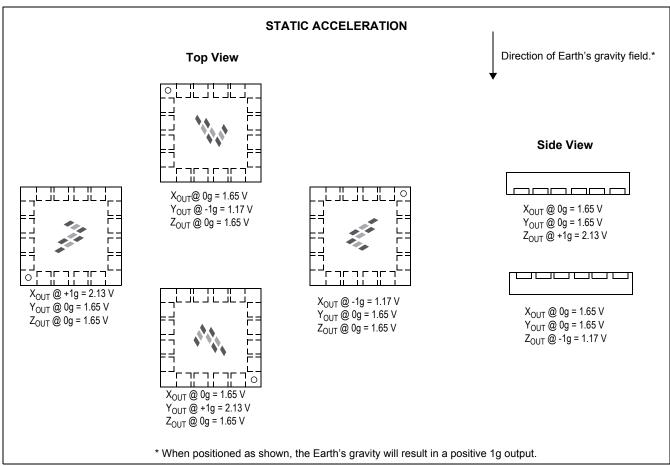
Figure 6. Recommended PCB Layout for Interfacing
Accelerometer to Microcontroller

NOTES:

- 1. Use 0.1 μ F capacitor on V_{DD} to decouple the power source. Do not exceed capacitor values of 2.2 or 3.3 μ F on V_{DD}-GND.
- 2. Physical coupling distance of the accelerometer to the microcontroller should be minimal.
- 3. Flag underneath package is connected to ground.
- 4. Place a ground plane beneath the accelerometer to reduce noise, the ground plane should be attached to all of the open ended terminals shown in Figure 6.
- 5. Use an RC filter with 1.0 k Ω and 0.1 μ F on the outputs of the accelerometer to minimize clock noise (from the switched capacitor filter circuit).
- 6. PCB layout of power and ground should not couple power supply noise.
- Accelerometer and microcontroller should not be a high current path.
- A/D sampling rate and any external power supply switching frequency should be selected such that they do not interfere with the internal accelerometer sampling frequency (11 kHz for the sampling frequency). This will prevent aliasing errors.

MMA7261Q

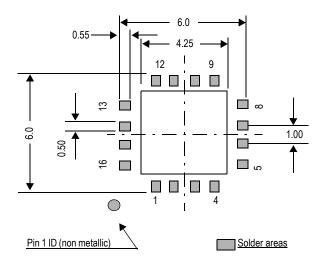




MINIMUM RECOMMENDED FOOTPRINT FOR SURFACE MOUNTED APPLICATIONS

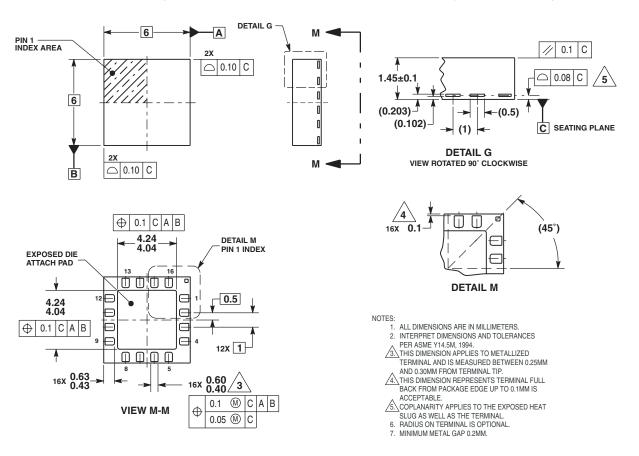
Surface mount board layout is a critical portion of the total design. The footprint for the surface mount packages must be the correct size to ensure proper solder connection interface between the board and the package.

With the correct footprint, the packages will self-align when subjected to a solder reflow process. It is always recommended to design boards with a solder mask layer to avoid bridging and shorting between solder pads.



PACKAGE DIMENSIONS

For the most current package revision, visit <u>www.freescale.com</u> and perform a keyword search using the "98A" listed below.



98ASA10651D ISSUE O 16-LEAD QFN

MMA7261Q

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